A new faster FFT approach using a novel algebra of subspace collections to computing the fields in composites

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Research Supported by the National Science Foundation

# Accelerating some Fast Fourier Transform Methods in two-component composites

The effective conductivity  $\sigma_*$  is an analytic function of the component conductivities  $\sigma_1$  and  $\sigma_2$ With  $\sigma_2 = 1$ ,  $\sigma_*(\sigma_1)$  has the properties of a Stieltjes function:  $\operatorname{Im}(\sigma_1)$ 

Bergman 1978 (pioneer, but faulty arguments) Milton 1981 (limit of resistor networks) Golden and Papanicolaou 1983 (rigorous proof)

## Original FFT approach of Moulinec and Suquet (1994,1998) based on the series expansion (Brown, Kroner, Willis...)

$$\sigma_* = \sigma_0 \mathbf{I} + \sum_{j=0}^{\infty} \Gamma_0 [\sigma(\mathbf{x}) - \sigma_0 \mathbf{I}] [\Gamma_1 (\mathbf{I} - \sigma/\sigma_0)]^j \Gamma_0, \quad \mathbf{e} = \mathbf{e}_0 + \sum_{j=0}^{\infty} [\Gamma_1 (\mathbf{I} - \sigma/\sigma_0)]^j \mathbf{e}_0,$$

 $\Gamma_0(\mathbf{k}) = \mathbf{I} \text{ if } \mathbf{k} = 0, \text{ zero otherwise.} \\ \Gamma_1(\mathbf{k}) = \mathbf{k} \mathbf{k}^T / (\mathbf{k} \cdot \mathbf{k}) \text{ for } \mathbf{k} \neq 0, \ \Gamma_1(\mathbf{0}) = 0$ 

Key point: the action of  $\Gamma_1$  is most easily evaluated in Fourier space, while the action of  $\sigma$  is most easily evaluated in real space. Therefore go back and forth between real and Fourier space, using FFTs, until the series converges.

With  $\sigma_0 = (\sigma_1 + \sigma_2)/2$  and  $\sigma_2 = 1$  one gets an expansion of the form

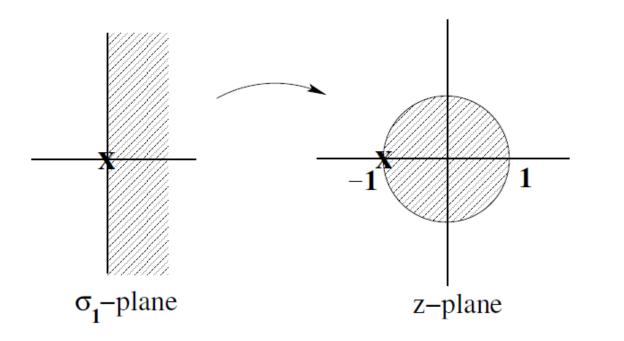
$$\sigma_*/\sigma_0 = 1 + \sum_{k=1}^{\infty} d_k \left(\frac{\sigma_1 - 1}{\sigma_1 + 1}\right), \qquad d_k = -\Gamma_0 [1 - 2\chi_1] \{\Gamma_1 [1 - 2\chi_1]\}^{k-1} \Gamma_0$$

**Complex analysis** provides the theory for the convergence of such expansions. The convergence and asymptotic rate of convergence is dictated by the nearest singularity to the origin in the  $(\sigma_1 - 1)/(\sigma_1 + 1)$ -plane.

### Numerical scheme of Moulinec and Suquet (1994)

$$\sigma_*/\sigma_0 = 1 + \sum_{k=1}^{\infty} d_k \left(\frac{\sigma_1 - 1}{\sigma_1 + 1}\right),$$

Let 
$$z = (\sigma_1 - 1)/(\sigma_1 + 1)$$



Taylor series converges in the unit disk in the z-plane, corresponding to the right-half plane in the  $\sigma_1$  plane.

Expect better convergence if there is no singularity near the origin.

**X** marks position of singularity, assumed here near the origin  $\sigma_1 = 0$ .

# Benchmark model example: a square array of squares at 25% volume fraction

Obnosov's exact formula

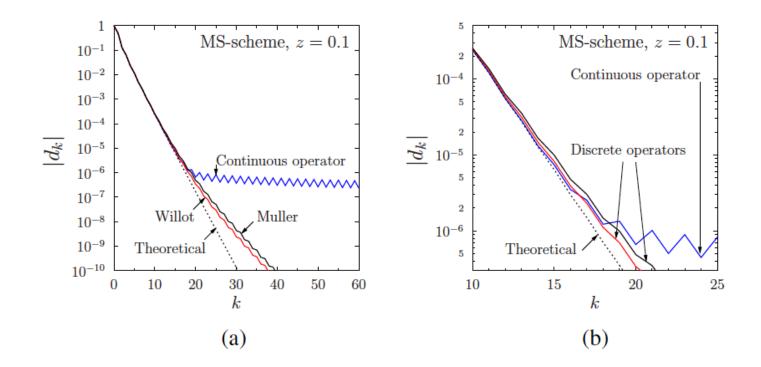
 $\sigma_* = \sqrt{(1+3\sigma_1)/(3+\sigma_1)}, \qquad \alpha = 1/3, \quad \beta = 3$ 

# Remark: there is also an exact formula for the effective conductivity of a 4-phase checkerboard

$$\lambda_1^* = \sqrt{\frac{\sigma_1 \sigma_2 \sigma_3 \sigma_4 (1/\sigma_1 + 1/\sigma_2 + 1/\sigma_3 + 1/\sigma_4)(\sigma_1 + \sigma_4)(\sigma_2 + \sigma_3)}{(\sigma_1 + \sigma_2 + \sigma_3 + \sigma_4)(\sigma_1 + \sigma_2)(\sigma_3 + \sigma_4)}},$$
  
$$\lambda_2^* = \sqrt{\frac{\sigma_1 \sigma_2 \sigma_3 \sigma_4 (1/\sigma_1 + 1/\sigma_2 + 1/\sigma_3 + 1/\sigma_4)(\sigma_1 + \sigma_2)(\sigma_3 + \sigma_4)}{(\sigma_1 + \sigma_2 + \sigma_3 + \sigma_4)(\sigma_1 + \sigma_4)(\sigma_2 + \sigma_3)}}.$$

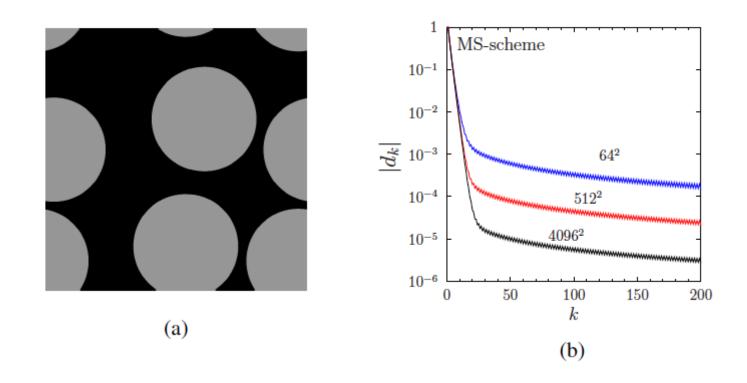
Conjectured by Mortola and Steffe (1985); proved indendently and by different approaches by Craster and Obnosov (2001) and Milton (2001)

### Series expansion coefficients:



Obnosov's microstructure, discretization  $512 \times 512$  pixels. MS scheme. Contrast z = 0.1. Coefficients  $d_k$  of the series (59) with different Green's operators: continuous operator (blue), Müller's operator [20] (black), Willot-Pellegrini operator [22] (red). (a): iterations 1 to 60. (b): close-up on iterations 10 to 25.

### Or for a more realistic geometry:



(a) Microstructure: 4 circular inclusions, volume fraction 50%. (b) Coefficients  $d_k$  of the series (55) at different resolutions. MS scheme.

#### Upshot: it does not make sense too iterate too many times

"Improved" FFT approach of Eyre and Milton (1999) based on the series expansion

$$oldsymbol{\sigma}_* = \sigma_0 \mathbf{I} + \sum_{j=0}^{\infty} \mathbf{\Gamma}_0 \mathbf{K} (\mathbf{\Upsilon} \mathbf{K})^j \mathbf{\Gamma}_0, \quad \mathbf{P} = \sum_{j=0}^{\infty} \mathbf{K} (\mathbf{\Upsilon} \mathbf{K})^j \mathbf{e}_0$$

$$\mathbf{K} = [\mathbf{I} + (\boldsymbol{\sigma} - \sigma_0 \mathbf{I})\mathbf{M}]^{-1}(\boldsymbol{\sigma} - \sigma_0 \mathbf{I}), \quad \boldsymbol{\Upsilon} = \mathbf{M} - (\boldsymbol{\Gamma}_1 / \sigma_0)$$

**M** is an arbitrary constant tensor, now  $\Upsilon$  can have eigenvalues of both signs.

Key point: the action of  $\Upsilon$  is most easily evaluated in Fourier space, while the action of **K** is most easily evaluated in real space. Therefore go back and forth between real and Fourier space, using FFTs, until the series converges.

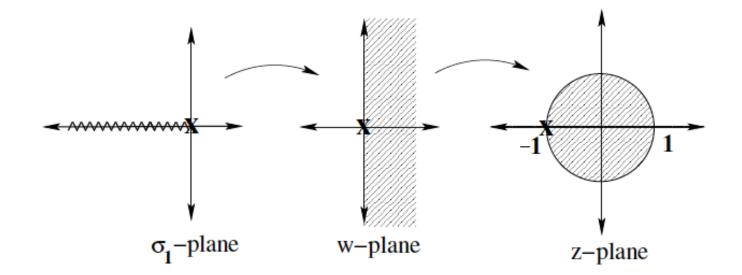
With 
$$\sigma_0 = \sqrt{\sigma_1 \sigma_2}$$
,  $\sigma_2 = 1$  and  $\mathbf{M} = \mathbf{I}/2\sigma_0$  one gets an expansion of the form

$$\sigma_*/\sqrt{\sigma_1} = 1 + \sum_{n=1}^{\infty} b_n \left(\frac{\sqrt{\sigma_1} - 1}{\sqrt{\sigma_1} + 1}\right)^n.$$

**Complex analysis** provides the theory for the convergence of such expansions. The convergence and asymptotic rate of convergence is dictated by the nearest singularity to the origin in the  $(\sqrt{\sigma_1} - 1)/(\sqrt{\sigma_1} + 1)$ -plane.

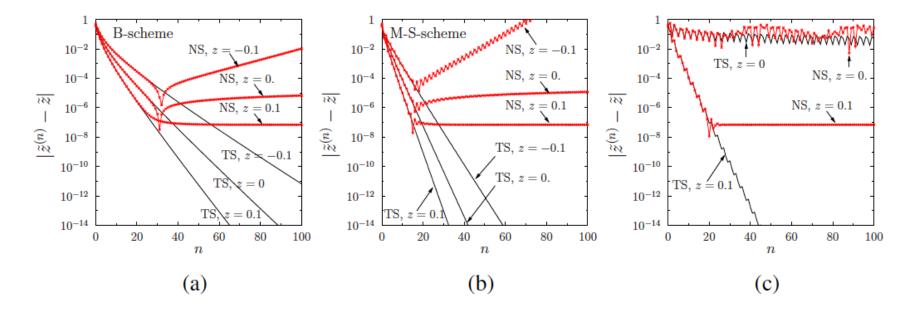
## Quick explanation of the "enhanced" rate of convergence of the Eyre-Milton Scheme

Let  $w = \sqrt{\sigma_1}$ , z = (w - 1)/(w + 1)



If we want a series expansion which converges in the entire the  $\sigma_1$ -plane minus the negative real  $\sigma_1$ -axis, then we first make a square root transformation which maps the cut complex  $\sigma_1$ -plane to the right half of the *w*-plane, followed by a fractional linear transformation which takes it to the unit disk in the *z*-plane, and find an expansion in powers of *z*. The scheme of Eyre and Milton (1999) provides such an expansion. Comparison of convergence for the Obnosov array of squares:

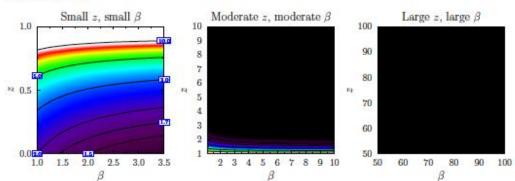
Here  $\tilde{z} = \sigma_*/\sigma_2$ , and  $z = \sigma_1/\sigma_2$ , TS=Theoretical Scheme, NS=Numerical Scheme



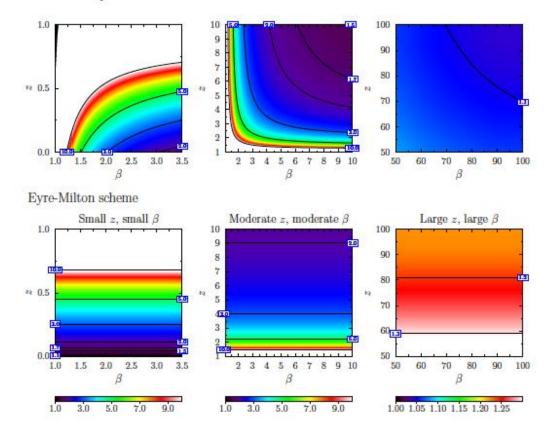
Obnosov problem. Comparison between the convergence of the theoretical series (TS) and the numerical series (NS) for different contrasts. (a) Reference medium: matrix (B-scheme). (b) Reference medium: arithmetic mean (MS-scheme). (c) Reference medium: geometric mean (EM-scheme). Discretization:  $512 \times 512$  pixels.

### None of the three schemes are entirely satisfactory

Brown scheme

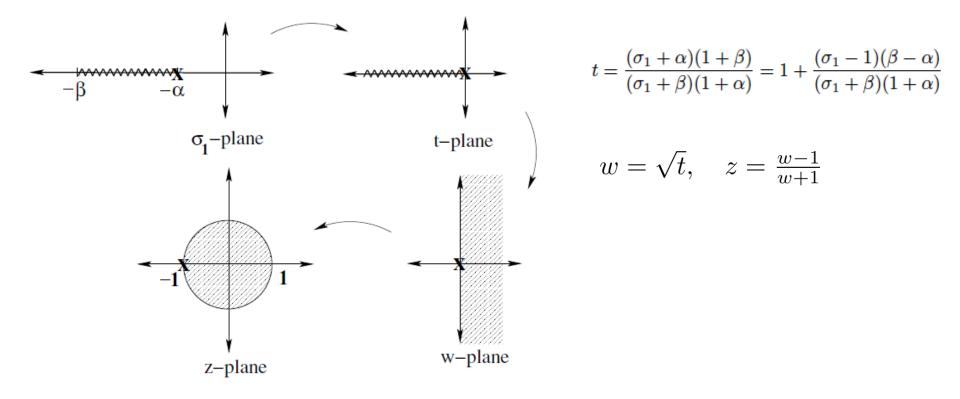


Moulinec-Suquet scheme



Snapshots of the rate of convergence for the 3 schemes in the plane  $(\beta, z)$ . The brighter the color, the faster the rate of convergence.

### If we know $\alpha, \beta$ then the ideal scheme should be:

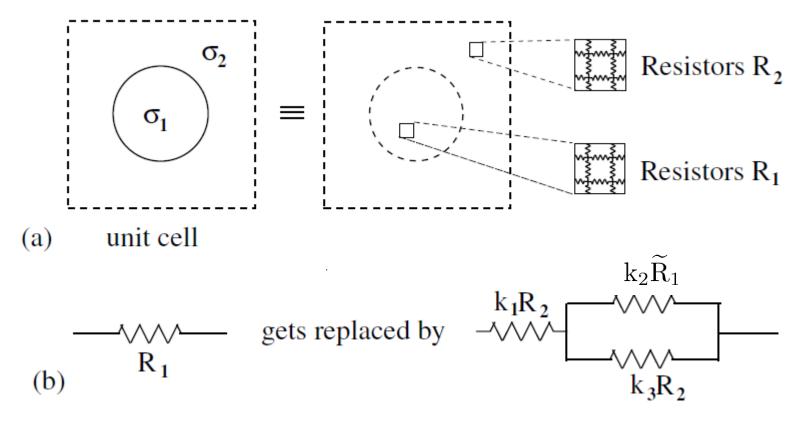


But we want to do this transformation at the level of the "subspace collection", to recover the fields We need a new series expansion for the fields.

## We need to find a FFT scheme that has an associated series expansion

$$\frac{\sigma_*}{\sqrt{[(\sigma_1 + \alpha)(1 + \beta)]/[(\sigma_1 + \beta)(1 + \alpha)]}} = 1 + \sum_{n=1}^{\infty} c_n \left(\frac{\sqrt{\frac{(\sigma_1 + \alpha)(1 + \beta)}{(\sigma_1 + \beta)(1 + \alpha)}} - 1}{\sqrt{\frac{(\sigma_1 + \alpha)(1 + \beta)}{(\sigma_1 + \beta)(1 + \alpha)}} + 1}\right)^n$$

### One idea: at a discrete level

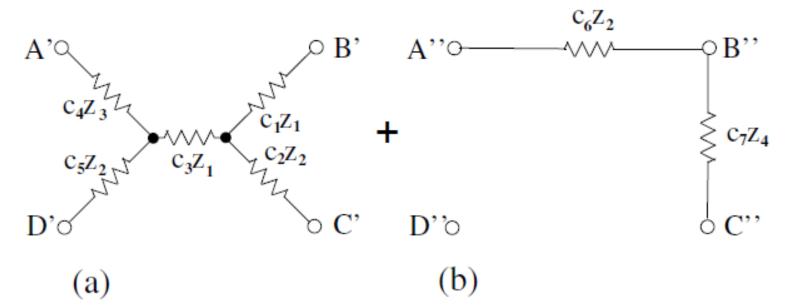


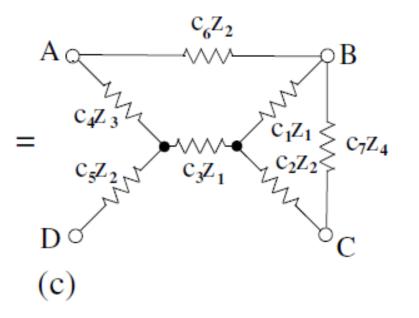
Locally, replace a 1-dimensional subspace by a 3-dimensional subspace

Problem: this substitution shortens the branch cut instead of lengthening it. Solution: Substitute "non-orthogonal subspace collections" Non-orthogonal subspace collections allow one to generalize the concept of function to

## Superfunctions!

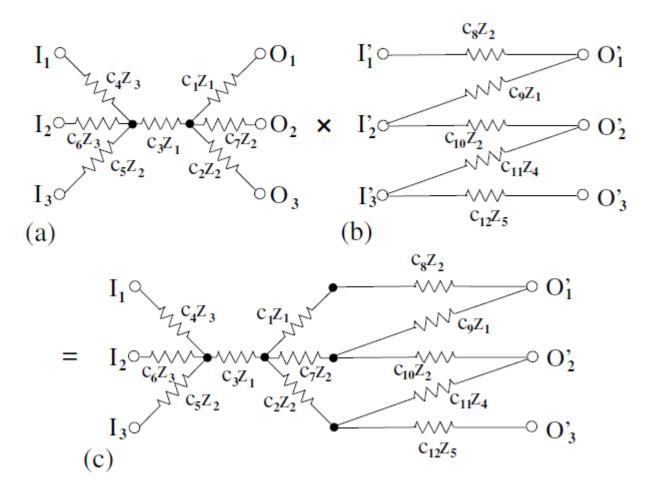
### Adding resistor networks





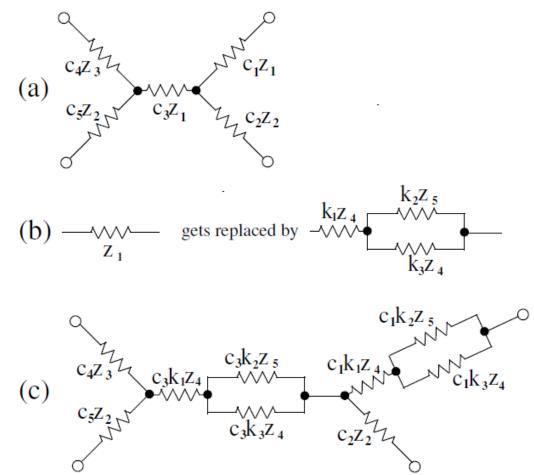
With non-orthogonal subspace collections one can subtract "resistor networks"

### Multiplying resistor networks



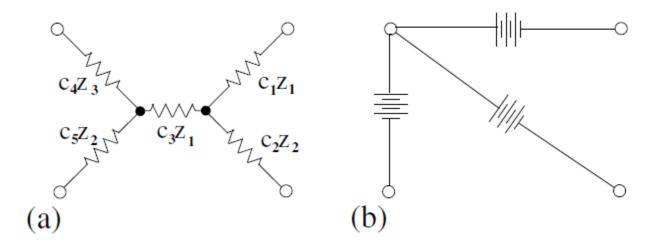
With non-orthogonal subspace collections one can divide "resistor networks".

### Substitution of networks



One is free to substitute a non-orthogonal subspace collection into an orthogonal one. This is precisely what we will do.

# We should consider a resistor network in conjunction with its batteries

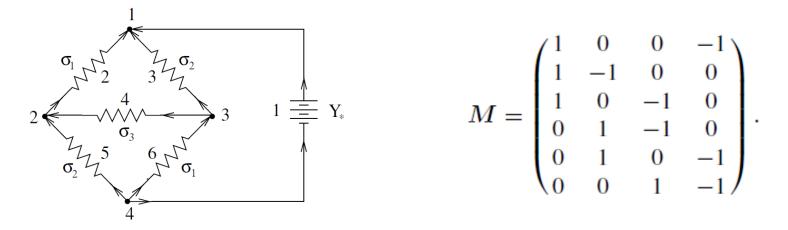


Space  $\mathcal{H}$ 

Space  $\mathcal{V}$ 

Combined Space  $\mathcal{K} = \mathcal{H} \oplus \mathcal{V}$ 

### **Incidence Matrices:**



 $M_{ij} = +1$  if the arrow of bond *i* points towards node *j*, = -1 if the arrow of bond *i* points away from node *j*,

= 0 if bond *i* and node *j* are not connected.

### Two natural subspaces:

- $\mathcal{J}$  the null space of  $M^{\top}$  (current vectors)
- ${\mathcal E}$  the range of M (potential drops)

These are orthogonal spaces and  $\mathcal{K} = \mathcal{E} \oplus \mathcal{J}$ 

#### Other spaces:

Divide the bonds in  $\mathcal{H}$  into n groups (representing the different impedances).

Define  $\mathcal{P}_i$  as the space of vectors  $\mathbf{P}$  with elements  $P_j$  that are zero if bond j is not in group i.

The projection  $\chi_i$  onto the space  $\mathcal{P}_i$  is diagonal and has elements

 $\{\chi_i\}_{jk} = 1$  if j = k and bond j is in group i,

= 0 otherwise.

Thus  $\mathcal{K} = \mathcal{E} \oplus \mathcal{J} = \mathcal{V} \oplus \mathcal{P}_1 \oplus \mathcal{P}_2 \oplus \cdots \oplus \mathcal{P}_n$ ,

This is an orthogonal subspace collection Y(n)

Abstract Theory of Composites; the Z(2)-problem

- Hilbert Space  $\mathcal{H} = \mathcal{U} \oplus \mathcal{E} \oplus \mathcal{J} = \mathcal{P}_1 \oplus \mathcal{P}_2$
- **Operator**  $\mathbf{L} = \sigma_1 \chi_1 + \sigma_2 \chi_2, \ \chi_i \text{ projects on } \mathcal{P}_i$
- Given  $\mathbf{E}_0 \in \mathcal{U}$
- Solve  $J_0 + J = L(E_0 + E)$
- With  $\mathbf{J}_0 \in \mathcal{U}$ ,  $\mathbf{J} \in \mathcal{J}$ ,  $\mathbf{E} \in \mathcal{E}$ ,
- Then  $\mathbf{J}_0 = \mathbf{L}_* \mathbf{E}_0$  defines  $\mathbf{L}_* : \mathcal{U} \to \mathcal{U}$

and  $\mathbf{L}_*$  is an analytic function of  $\sigma_1$  and  $\sigma_2$ ,  $\mathbf{L}_*(\sigma_1, \sigma_2)$ 

Example: 2-Phase Conducting Composites

- ${\cal H}$  Periodic fields that are square integrable over the unit cell
- $\mathcal{U}$  Constant vector fields (the "applied fields")
- ${\mathcal E}$  Gradients of periodic potentials
- ${\cal J}$  Fields with zero divergence and zero average value
- $\mathcal{P}_i$  Fields that are non-zero only in phase i
- $\mathbf{E}_0 + \mathbf{E}(\mathbf{x})$  Total electric field
- $\mathbf{J}_0 + \mathbf{J}(\mathbf{x})$  Total current field
- $\mathbf{L}=\pmb{\sigma}(\mathbf{x})$  Local conductivity
- $\mathbf{L}_* = oldsymbol{\sigma}_*$  Effective conductivity

**Abstract Theory of Composites; the Y(2)-problem** 

- Hilbert Space  $\mathcal{K} = \mathcal{E} \oplus \mathcal{J} = \mathcal{V} \oplus \mathcal{H}, \quad \mathcal{H} = \mathcal{P}_1 \oplus \mathcal{P}_2$
- Operator  $\mathbf{L} = \sigma_1 \chi_1 + \sigma_2 \chi_2, \ \chi_i \text{ projects on } \mathcal{P}_i$
- Given  $\mathbf{E}_0 \in \mathcal{V}$
- Solve  $\mathbf{J}_1 = \mathbf{L}\mathbf{E}_1$
- With  $\mathbf{J}_1, \mathbf{E}_1 \in \mathcal{H}, \quad \mathbf{J}_0 + \mathbf{J}_1 \in \mathcal{J}, \quad \mathbf{E}_0 + \mathbf{E}_1 \in \mathcal{E}$
- Then  $\mathbf{J}_0 = -\mathbf{Y}_*\mathbf{E}_0$  defines  $\mathbf{Y}_* : \mathcal{V} \to \mathcal{V}$

and  $\mathbf{Y}_*$  is an analytic function of  $\sigma_1$  and  $\sigma_2$ ,  $\mathbf{Y}_*(\sigma_1, \sigma_2)$ 

Y(n) subspace collection:

 $\mathcal{K} = \mathcal{E} \oplus \mathcal{J} = \mathcal{V} \oplus \mathcal{P}_1 \oplus \mathcal{P}_2 \oplus \cdots \oplus \mathcal{P}_n,$ 

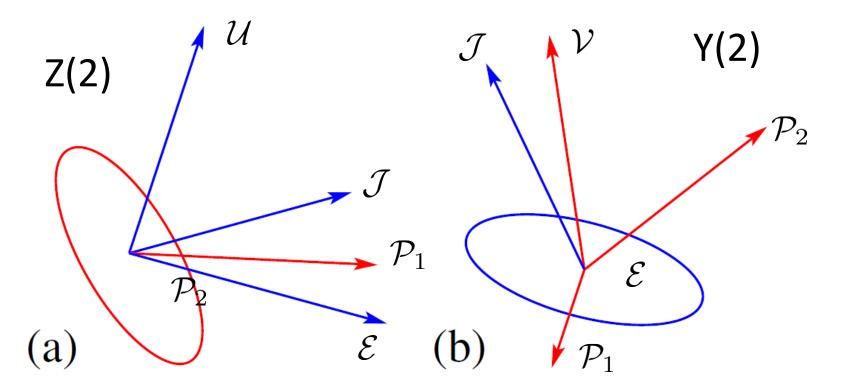
Z(n) subspace collection:

 $\mathcal{H} = \mathcal{U} \oplus \mathcal{E} \oplus \mathcal{J} = \mathcal{P}_1 \oplus \mathcal{P}_2 \oplus \cdots \oplus \mathcal{P}_n,$ 

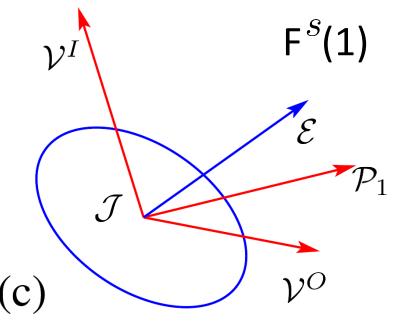
Superfunction F<sup>s</sup>(n): Y(n) subspace collection with

 $\mathcal{V} = \mathcal{V}^I \oplus \mathcal{V}^O$ 

"Subspace collections" need not have orthogonal subspaces



Key: Allow nonorthogonal "subspace collections". Then we have a whole algebra: can define "subtraction" and "division" of subspace collections.



### The vector subspace collection that we substitute into the original subspace collection

Want an analog of replacing  $- \underset{R_1}{\overset{k_1R_2}{\longrightarrow}}$  with  $- \underset{k_3R_2}{\overset{k_2R_1}{\longrightarrow}}$  using non-orthogonal subspaces

Now consider a 3-dimensional subspace collection  $\mathcal{H}'$  consisting of 3 component vectors  $\mathbf{P} = [P_1, P_2, P_3]^T$ with inner product

$$(\mathbf{P}, \widetilde{\mathbf{P}}) = \sum_{i=1}^{3} \overline{P}_i \widetilde{P}_i,$$

where the overline denotes complex conjugation. The projection  $\chi' = p \otimes p$  projects onto the one dimensional space of fields proportional to the unit vector  $\mathbf{p}$  where  $\mathbf{p} = [p_1, p_2, p_3]^T$  and  $p_1, p_2, p_3$  are given constants such that  $p_1^2 + p_2^2 + p_3^2 = 1$ . The *p*'s could be complex but we *do not* mean  $|p_1|^2 + |p_2|^2 + |p_3|^2 = 1$ . Thus  $\chi'$  is a projection but not an orthogonal projection when the *p*'s are complex, as then  $\chi' = p \otimes p$  is not Hermitian. We take the following:

 $\mathcal{U}'$  is the space of fields proportional to  $(1,0,0)^T$ ,  $\mathcal{E}'$  is the space of fields proportional to  $(0,1,0)^T$ ,  $\mathcal{J}'$  is the space of fields proportional to  $(0,0,1)^T$ ,  $\mathcal{P}_1$  is the space of fields proportional to  $(p_1, p_2, p_3)^T$ ,  $\mathcal{P}_2$  is the space of fields  $(P_1, P_2, P_3)^T$ such that  $p_1P_1 + p_2P_2 + p_3P_3 = 0$ . The field equations become

$$\mathbf{J}' = [(t - \sigma_2)\chi' + \sigma_2]\mathbf{E}', \quad \mathbf{E}' \in \mathcal{U}' \oplus \mathcal{E}', \quad \mathbf{J}' \in \mathcal{U}' \oplus \mathcal{J}',$$

where the constant t will be chosen so the associated "effective modulus" is  $\sigma_1$ . That is

$$\Gamma_0 \mathbf{J}' = \sigma_1 \Gamma_0 \mathbf{E}',$$

where  $\Gamma_0$  is the projection onto  $\mathcal{U}$ , so that

$$J_1' = \sigma_1 E_1'.$$

Without loss of generality we can choose  $E'_1 = 1, J'_1 = \sigma_1$  so the field equations become

$$\begin{pmatrix} J_1' \\ 0 \\ J_3' \end{pmatrix} = (t - \sigma_2) \underbrace{\begin{pmatrix} p_1^2 & p_1 p_2 & p_1 p_3 \\ p_1 p_2 & p_2^2 & p_2 p_3 \\ p_1 p_3 & p_2 p_3 & p_3^2 \end{pmatrix}}_{\chi'} \begin{pmatrix} E_1' \\ E_2' \\ 0 \end{pmatrix} + \sigma_2 \begin{pmatrix} E_1' \\ E_2' \\ 0 \end{pmatrix}.$$

From the middle equation we get

$$(t - \sigma_2)p_1p_2E'_1 + [(t - \sigma_2)p_2^2 + \sigma_2]E'_2 = 0,$$

which with  $E'_1 = 1$  gives

$$E_2' = \frac{(\sigma_2 - t)p_1p_2}{(t - \sigma_2)p_2^2 + \sigma_2}.$$

So we have

$$\begin{split} \sigma_1 &= J_1' &= p_1^2(t - \sigma_2) + \sigma_2 - \frac{(\sigma_2 - t)^2 p_1^2 p_2^2}{(t - \sigma_2) p_2^2 + \sigma_2} \\ &= \sigma_2 + \frac{p_1^2 \sigma_2 (t - \sigma_2)}{(t - \sigma_2) p_2^2 + \sigma_2} \\ &= \sigma_2 + \frac{p_1^2 \sigma_2}{p_2^2 + \sigma_2 / (t - \sigma_2)}, \end{split}$$

which with  $\sigma_2 = 1$  is satisfied with

$$t = 1 + \frac{\sigma_1 - 1}{p_1^2 - p_2^2(\sigma_1 - 1)} = 1 + \frac{(\sigma_1 - 1)(\beta - \alpha)}{(\sigma_1 + \beta)(1 + \alpha)},$$

where

$$\begin{array}{rcl} \alpha & = & -1 - \frac{p_1^2}{p_2^2 - 1}, \\ \\ \beta & = & -1 - \frac{p_1^2}{p_2^2}. \end{array}$$

The parameters  $p_1$  and  $p_2$  need to be complex.

#### The Hilbert space after the substitution

Now consider the Hilbert space  $\mathcal{H}''$  consisting of all periodic fields of the form

$$\mathbf{P}''(\mathbf{x}) = \underbrace{\begin{pmatrix} \mathbf{0} \\ \mathbf{S}(\mathbf{x}) \\ \mathbf{T}(\mathbf{x}) \end{pmatrix}}_{\in \mathcal{P}_1 \otimes (\mathcal{E}' \oplus \mathcal{J}')} \chi(\mathbf{x}) + \underbrace{\begin{pmatrix} \mathbf{Q}(\mathbf{x}) \\ \mathbf{0} \\ \mathbf{0} \\ \\ \in \mathcal{H} \otimes \mathcal{U}' \end{bmatrix}}_{\in \mathcal{H} \otimes \mathcal{U}'} = \begin{pmatrix} \mathbf{Q}(\mathbf{x}) \\ \chi(\mathbf{x}) \mathbf{S}(\mathbf{x}) \\ \chi(\mathbf{x}) \mathbf{T}(\mathbf{x}) \end{pmatrix}.$$

Fields in  $\mathcal{U}''$  take the form

$$\mathbf{u}''(\mathbf{x}) = \begin{pmatrix} \mathbf{u}_0 \\ 0 \\ 0 \end{pmatrix} \in \mathcal{U} \otimes \mathcal{U}'.$$

Fields in  $\mathcal{E}''$  take the form

$$\mathbf{E}''(\mathbf{x}) = \underbrace{\begin{pmatrix} \mathbf{0} \\ \mathbf{S}(\mathbf{x}) \\ \mathbf{0} \end{pmatrix}}_{\in \mathcal{P}_1 \otimes \mathcal{E}'} \chi(\mathbf{x}) + \underbrace{\begin{pmatrix} \widetilde{\mathbf{E}}(\mathbf{x}) \\ \mathbf{0} \\ \mathbf{0} \end{pmatrix}}_{\in \mathcal{E} \otimes \mathcal{U}'} = \begin{pmatrix} \widetilde{\mathbf{E}}(\mathbf{x}) \\ \mathbf{S}(\mathbf{x}) \chi(\mathbf{x}) \\ \mathbf{0} \end{pmatrix},$$

where  $\widetilde{\mathbf{E}}(\mathbf{x}) \in \mathcal{E}$ . Fields in  $\mathcal{J}''$  take the form

$$\mathbf{J}''(\mathbf{x}) = \underbrace{\begin{pmatrix} 0 \\ 0 \\ \mathbf{T}(\mathbf{x}) \end{pmatrix}}_{\in \mathcal{P}_1 \otimes \mathcal{J}'} \chi(\mathbf{x}) + \underbrace{\begin{pmatrix} \widetilde{\mathbf{J}}(\mathbf{x}) \\ 0 \\ 0 \\ \in \mathcal{J} \otimes \mathcal{U}' \end{pmatrix}}_{\in \mathcal{J} \otimes \mathcal{U}'} = \begin{pmatrix} \widetilde{\mathbf{J}}(\mathbf{x}) \\ 0 \\ \mathbf{T}(\mathbf{x}) \chi(\mathbf{x}) \end{pmatrix}, \quad \text{ where } \widetilde{\mathbf{J}}(\mathbf{x}) \in \mathcal{J}.$$

The space  $\mathcal{P}'_1$  consists of all vectors of the form

$$c \begin{pmatrix} p_1 \\ p_2 \\ p_3 \end{pmatrix},$$

and  $\mathcal{P}_1''$  consists of all fields  $\mathbf{P}(\mathbf{x})$  of the form

$$\begin{pmatrix} p_1 \mathbf{C}(\mathbf{x}) \ p_2 \mathbf{C}(\mathbf{x}) \ p_3 \mathbf{C}(\mathbf{x}) \end{pmatrix} \chi(\mathbf{x}) \in \mathcal{P}_1 \otimes \mathcal{P}_1'.$$

Also  $\mathcal{P}'_2$  consists of all vectors of the form

$$c\begin{pmatrix} q_1\\ q_2\\ q_3 \end{pmatrix}$$
 where  $p_1q_1 + p_2q_2 + p_3q_3 = 0$ ,

and  $\mathcal{P}_2''$  consists of all fields  $\mathbf{P}(\mathbf{x})$  of the form

$$\underbrace{\begin{pmatrix} \mathbf{Q}_1(\mathbf{x}) \\ \mathbf{Q}_2(\mathbf{x}) \\ \mathbf{Q}_3(\mathbf{x}) \end{pmatrix} \chi(\mathbf{x}) + (1 - \chi(\mathbf{x})) \begin{pmatrix} \mathbf{R}(\mathbf{x}) \\ 0 \\ 0 \end{pmatrix}}_{\in (\mathcal{P}_1 \otimes \mathcal{P}'_2 + \mathcal{P}_2 \otimes \mathcal{U}')}$$

where  $p_1 \mathbf{Q}_1(\mathbf{x}) + p_2 \mathbf{Q}_2(\mathbf{x}) + p_3 \mathbf{Q}_3(\mathbf{x}) = 0.$ 

The inner product on  $\mathcal{H}''$  is defined to be

$$(\mathbf{P}, \widetilde{\mathbf{P}}) = \int_{\text{unit cell}} [\overline{\mathbf{S}(\mathbf{x})} \cdot \widetilde{\mathbf{S}}(\mathbf{x}) + \overline{\mathbf{T}(\mathbf{x})} \cdot \widetilde{\mathbf{T}}(\mathbf{x})] \boldsymbol{\chi}(\mathbf{x}) + \overline{\mathbf{Q}(\mathbf{x})} \cdot \widetilde{\mathbf{Q}}(\mathbf{x}),$$

We define  $\chi'' = (\mathbf{p} \otimes \mathbf{p}) \chi$ , i.e.,

$$\chi'' \left\{ \begin{pmatrix} 0\\ \mathbf{S}(\mathbf{x})\\ \mathbf{T}(\mathbf{x}) \end{pmatrix} \chi(\mathbf{x}) + \begin{pmatrix} \mathbf{Q}(\mathbf{x})\\ 0\\ 0 \end{pmatrix} \right\} = \begin{pmatrix} p_1^2 \mathbf{I} & p_1 p_2 \mathbf{I} & p_1 p_3 \mathbf{I} \\ p_1 p_2 \mathbf{I} & p_2^2 \mathbf{I} & p_2 p_3 \mathbf{I} \\ p_1 p_3 \mathbf{I} & p_2 p_3 \mathbf{I} & p_3^2 \mathbf{I} \end{pmatrix} \begin{pmatrix} \mathbf{Q}(\mathbf{x})\\ \mathbf{S}(\mathbf{x})\\ \mathbf{T}(\mathbf{x}) \end{pmatrix} \chi(\mathbf{x}),$$

The field equations become

$$\mathbf{J}'' = [(t - \sigma_2)\chi'' + \sigma_2 \mathbf{I}]\mathbf{E}'', \quad \mathbf{E}'' \in \mathcal{U}'' \oplus \mathcal{E}'', \quad \mathbf{J}'' \in \mathcal{U}'' \oplus \mathcal{J}''.$$

These are easy to solve given periodic solutions J(x) and E(x) to the equations in the Hilbert space  $\mathcal{H}$ , i.e.,

$$\mathbf{J} = [(\sigma_1 - \sigma_2)\boldsymbol{\chi} + \sigma_2]\mathbf{E}, \quad \nabla \cdot \mathbf{J} = 0, \quad \nabla \times \mathbf{E} = 0.$$

We take (with  $E'_1 = 1$ )

$$\mathbf{E}'' = \begin{pmatrix} \mathbf{E}(\mathbf{x}) \\ E'_2 \mathbf{E}(\mathbf{x}) \boldsymbol{\chi}(\mathbf{x}) \\ 0 \end{pmatrix}, \quad \mathbf{J}'' = \begin{pmatrix} \mathbf{J}(\mathbf{x}) \\ 0 \\ J'_3 \mathbf{J}(\mathbf{x}) / \sigma_1 \end{pmatrix}.$$

Note that we have  $\mathbf{E}'' \in \mathcal{U}'' \oplus \mathcal{E}''$  and  $\mathbf{J}'' \in \mathcal{U}'' \oplus \mathcal{J}''$ . Also, with  $\sigma_2 = 1$ , we have

$$\begin{aligned} ((t-\sigma_2)\chi''+\sigma_2)\mathbf{E}'' &= (\mathbf{p}\otimes\mathbf{p}(t-\sigma_2)+\sigma_2])\begin{pmatrix} 1\\ E_2'\mathbf{E}(\mathbf{x})\chi(\mathbf{x})\\ 0 \end{pmatrix} \\ &+ \begin{pmatrix} \mathbf{E}(\mathbf{x})(1-\chi(\mathbf{x}))\\ 0\\ 0 \end{pmatrix} \\ &= \begin{pmatrix} J_1'\mathbf{E}(\mathbf{x})\\ 0\\ J_3'\mathbf{E}(\mathbf{x}) \end{pmatrix}\chi(\mathbf{x}) + \begin{pmatrix} \mathbf{E}(\mathbf{x})\\ 0\\ 0 \end{pmatrix}(1-\chi(\mathbf{x})) \\ &= \begin{pmatrix} \mathbf{J}(\mathbf{x})\\ 0\\ J_3'\mathbf{E}(\mathbf{x}) \end{pmatrix}\chi(\mathbf{x}) + (1-\chi(\mathbf{x}))\begin{pmatrix} \mathbf{J}(\mathbf{x})\\ 0\\ 0 \end{pmatrix} = \mathbf{J}''. \end{aligned}$$

Finally if  $\Gamma_0''$  is the projection onto  $\mathcal{U}''$  we have

$$\begin{split} \mathbf{\Gamma}_0'' \mathbf{E}'' &= \begin{pmatrix} \langle \mathbf{E} \rangle \\ 0 \\ 0 \end{pmatrix} \boldsymbol{\chi}(\mathbf{x}) + \begin{pmatrix} \langle \mathbf{E} \\ 0 \\ 0 \end{pmatrix} \rangle (1 - \boldsymbol{\chi}(\mathbf{x})), \\ \mathbf{\Gamma}_0'' \mathbf{J}'' &= \begin{pmatrix} \langle \mathbf{J} \rangle \\ 0 \\ 0 \end{pmatrix} \boldsymbol{\chi}(\mathbf{x}) + \begin{pmatrix} \langle \mathbf{J} \rangle \\ 0 \\ 0 \end{pmatrix} (1 - \boldsymbol{\chi}(\mathbf{x})), \end{split}$$

and since  $\langle \mathbf{J} \rangle = \sigma_* \langle \mathbf{E} \rangle$  we deduce that

$$\mathbf{\Gamma}_0''\mathbf{J}'' = \sigma_*\mathbf{\Gamma}_0''\mathbf{E}''.$$

#### UPSHOT: $\sigma_*$ is still the effective tensor.

As expected, the Hilbert space substitution did not change it, but it does change the convergence rate of series expansions. The operator  $\chi''$  is easily evaluated in real space. The operator  $\Gamma''_1$  which projects onto  $\mathcal{E}''$  is easily evaluated in Fourier space since

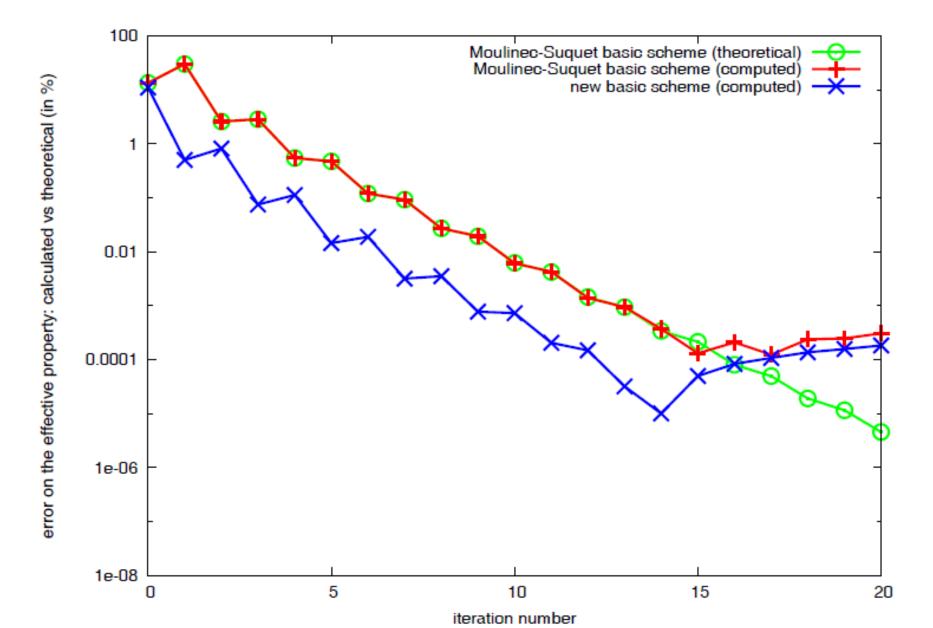
$$\Gamma_1'' \begin{pmatrix} \mathbf{Q}(\mathbf{x}) \\ \chi(\mathbf{x}) \mathbf{S}(\mathbf{x}) \\ \chi(\mathbf{x}) \mathbf{T}(\mathbf{x}) \end{pmatrix} = \begin{pmatrix} \Gamma_1 \mathbf{Q}(\mathbf{x}) \\ \chi(\mathbf{x}) \mathbf{S}(\mathbf{x}) \\ 0 \end{pmatrix}, \qquad (8.46)$$

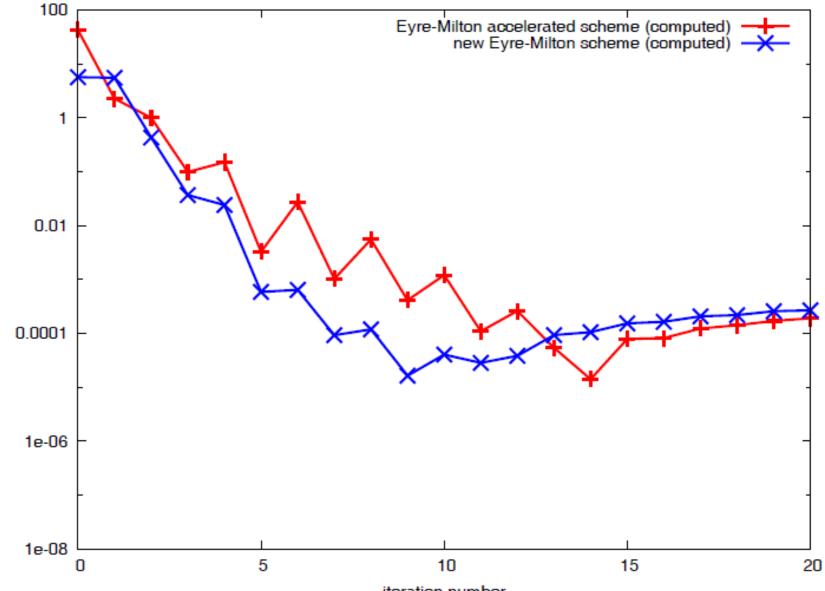
where in Fourier space

$$\Gamma_1 \widehat{\mathbf{Q}}(\mathbf{k}) = \begin{cases} \frac{\mathbf{k} \otimes \mathbf{k} \widehat{\mathbf{Q}}(\mathbf{k})}{|\mathbf{k}|^2}, & \mathbf{k} \neq 0, \\ 0, & \mathbf{k} = 0. \end{cases}$$
(8.47)

Hence the Fast Fourier Transform methods of Moulinec and Suquet and of Milton and Eyre can be diectly applied in the Hilbert soace  $\mathcal{H}$ ".

### Does the idea work? YES!





iteration number

Remark: The analysis relied heavily on knowledge of the parameters  $\alpha$  and  $\beta$  that are associated with the spectrum of the operator  $\Gamma_1 \chi_1 \Gamma_1$ . For periodic arrays of disks or spheres, bounds on the spectrum have been obtained by Bruno (1991). For more general geometries, and for elasticity and other problems, a new approach to getting bounds on the spectrum can be found here:

JOURNAL OF MATHEMATICAL PHYSICS 59, 061508 (2018)

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### Thank you!

Thank you!

Thank you!





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